



# SPE06S60F-A

## 主要参数 MAIN CHARACTERISTICS

600V/6A IGBT 三相全桥智能功率模块	
V <sub>CES</sub>	600V
± I <sub>C</sub>	6A
± I <sub>CP</sub>	12A

## 用途

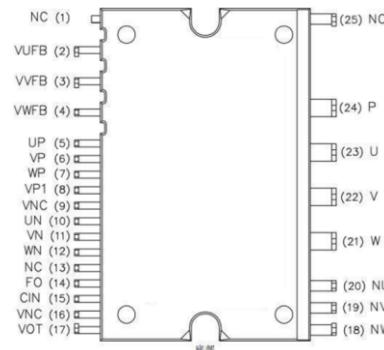
- 冰箱压缩机
  - 油烟机
  - 风扇
  - 空气净化器
  - 洗碗机水泵
- Refrigerator compressor
  - Lampblack machine
  - Electric fan
  - Air purifier
  - Dishwasher pump

## APPLICATIONS

## 封装 Package



## DIP25-FP



## PIN1-PIN25

## 产品特性

- 下臂 IGBT 发射极输出，内置自举二极管。
- 600V/6A 三相逆变器，内置低损耗沟道栅-场截止型 IGBT。
- IGBT 驱动：增强型输入滤波，上下臂互锁，高速 600V 电平转换，电源欠压保护，短路（过流）保护。
- 故障信号：对应于短路（过流）和 VP1 电源欠压故障。
- 输入接口：兼容 3.3V & 5V 输入信号，高电平有效。
- 温度检测：负温度系数热敏电阻检测输出。

## FEATURES

- Lower arm IGBT emitter output, built-in bootstrap diode.
- 600V/6A three-phase inverter with built-in low-loss trench gate-field stop IGBT.
- IGBT drive: enhanced input filtering, interlocking of upper and lower arms, high-speed 600V level conversion, power supply undervoltage protection, short circuit (overcurrent) protection.
- Fault signal: corresponds to short circuit (overcurrent) and VP1 power undervoltage fault.
- Input interface: compatible with 3.3V&5V input signals, high level effective.
- Temperature detection: negative temperature coefficient thermistor detection output.

## 订货信息 ORDER MESSAGE

订 购 料 号 Order number	产 品 信 息 Product information			
	无卤-条管 Halogen-Free-Tube	无卤-编带 Halogen-Free-Reel	印 记 Marking	封 装 Package
2A01-0355-16	SPE06S60F-A	N/A	SPE06S60F-A	DIP25-FP

## 模块示意图 Module diagram

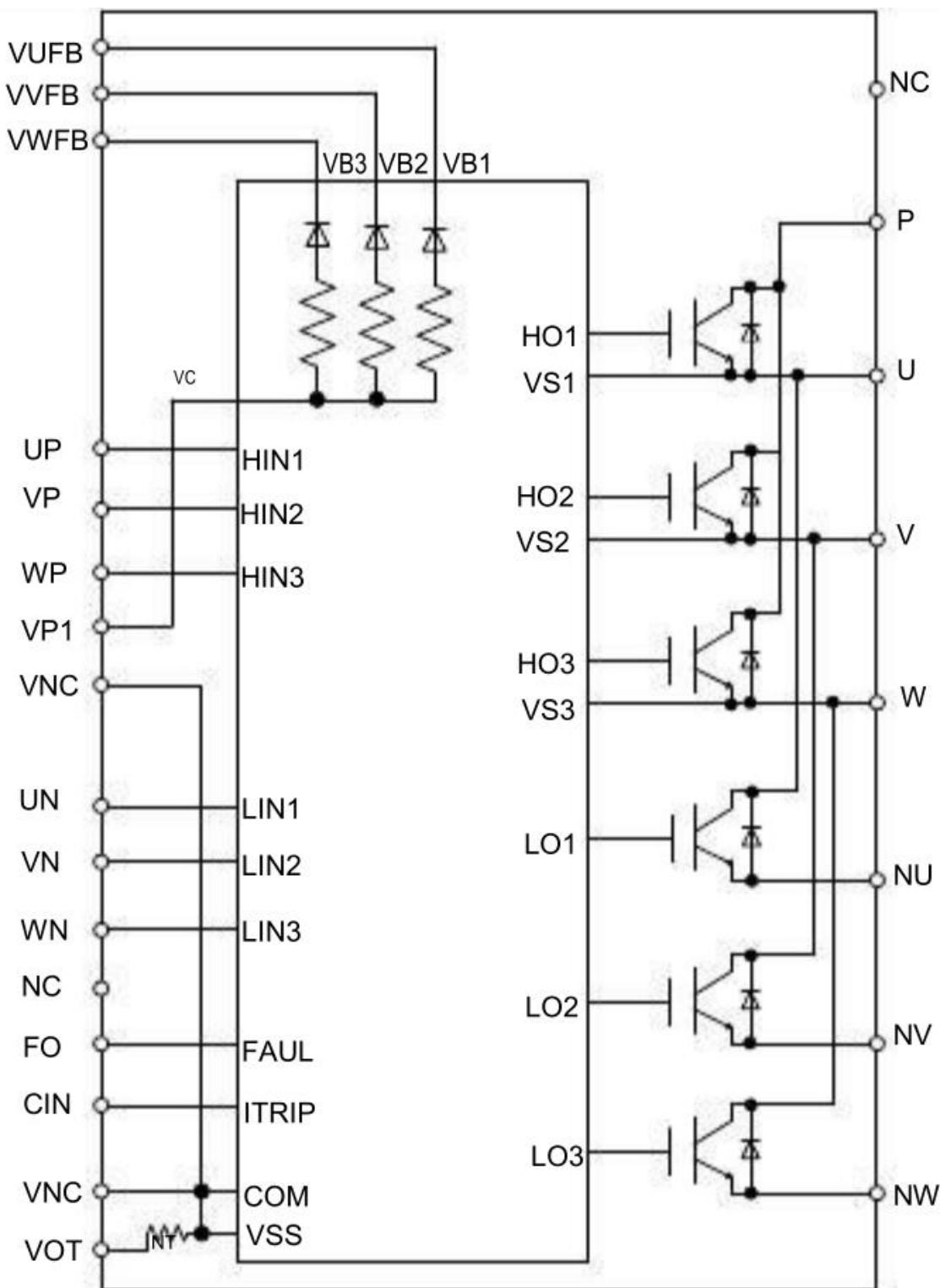


图 1：模块内部电路示意图

Fig 1: Internal circuit



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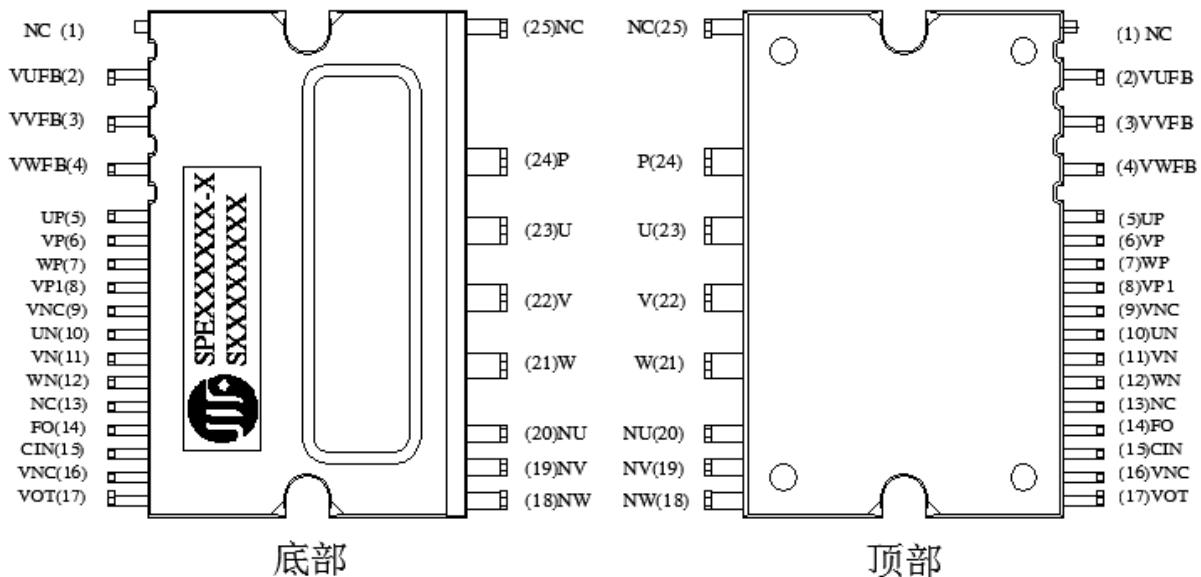


图 2: 模块引脚分布示意图

Fig 2: Distribution of pin

管脚编号 Pin Number	管脚名称 Pin Names	管脚描述 Pin Description
1	NC	无连接 connectionless
2	VUFB	U 相上臂驱动电源端子 U-phase upper arm drive power terminal
3	VVFB	V 相上臂驱动电源端子 V phase upper arm drive power terminal
4	VWFB	W 相上臂驱动电源端子 W phase upper arm drive power terminal
5	UP	U 相上臂控制信号输入端子 U-phase upper arm control signal input terminal
6	VP	V 相上臂控制信号输入端子 V phase upper arm control signal input terminal
7	WP	W 相上臂控制信号输入端子 W phase upper arm control signal input terminal
8	VP1	控制电源端子 Control power terminal
9	VNC	控制电源 GND 端子 Control power GND terminal
10	UN	U 相下臂控制信号输入端子 U-phase lower arm control signal input terminal
11	VN	V 相下臂控制信号输入端子 V phase lower arm control signal input terminal
12	WN	W 相下臂控制信号输入端子 W phase lower arm control signal input terminal
13	NC	无连接 connectionless
14	FO	故障输出端子 Fault output terminal
15	CIN	短路保护触发电压检测端子 Short circuit protection trigger voltage detection terminal



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16	VNC	控制电源 GND 端子 Control power GND terminal
17	VOT	温度检测输出端子 Temperature detection output terminal
18	NW	W 相下臂 IGBT 发射极端子 W phase lower arm IGBT emitter terminal
19	NV	V 相下臂 IGBT 发射极端子 V phase lower arm IGBT emitter terminal
20	NU	U 相下臂 IGBT 发射极端子 U phase lower arm IGBT emitter terminal
21	W	W 相输出端子 W phase output terminal
22	V	V 相输出端子 V phase output terminal
23	U	U 相输出端子 U phase output terminal
24	P	逆变器直流输入端子 Dc input terminal of inverter
25	NC	无连接 connectionless

图 3: 模块引脚功能定义表

Fig 3: Pin function

最大额定值 ( $T_j = 25^\circ\text{C}$ ,除非特殊说明)**Absolute Maximum Ratings** ( $T_j = 25^\circ\text{C}$ , Unless otherwise Specified)

## 逆变部分 Inverter Part

项号 Mark	项目 Project	条件 Condition	额定值 Ratings	单位 Units
$V_{cc}$	电源电压 Supply voltage	应用于 P- NU, NV, NW 之间 It applies between p-nu, NV, and NW	450	V
$V_{CC(Surge)}$	电源电压 (含浪涌) Supply voltage (including surge)	应用于 P- NU, NV, NW 之间 It applies between p-nu, NV, and NW	500	V
$V_{CES}$	集电极-发射极间电压 Collector to emitter voltage	/	600	V
$\pm I_c$	集电极电流 Collector current	$T_c = 25^\circ\text{C}$ ( $T_c$ 测试方法见图 4) $TC = 25^\circ\text{C}$ (see Figure 4 for $T_c$ test method)	6	A
$\pm I_{CP}$	集电极电流 (峰值) Collector current (peak)	$T_c = 25^\circ\text{C}$ , 脉冲宽度小于 1ms $TC = 25^\circ\text{C}$ , pulse width is less than 1ms	12	A
$P_c$	集电极功耗 Collector power consumption	$T_c = 25^\circ\text{C}$ , 单晶片 $TC = 25^\circ\text{C}$ , single chip	23	W
$T_j$	结温 Junction temperature	(见备注 1) (See note 1)	-40~+150	°C

备注 1: IPM 功率晶片最大额定结温为  $150^\circ\text{C}$ (@表面温度  $T_c \leq 100^\circ\text{C}$ )。然而,为了确保 IPM 运行安全, 结温应限定于  $T_j(av) \leq 125^\circ\text{C}$  (@表面温度  $T_c \leq 100^\circ\text{C}$ )。

Remark 1: The maximum rated junction temperature of the IPM power chip is  $150^\circ\text{C}$  (@surface temperature  $TC \leq 100^\circ\text{C}$ ). However, to ensure safe operation of the IPM, the junction temperature should be limited to  $T_j(av) \leq 125^\circ\text{C}$  (@surface



temperature  $T_c \leq 100^\circ\text{C}$ ).

### 控制部分 Control Part

记号 Mark	项目 Project	条件 Condition	额定值 Ratings	单位 Units
VDB	上桥臂控制电源电压 Upper arm control supply voltage	应用于 UFB – U, VFB-V, WFB-W 之间 Applied between UFB-U, VFB-V, WFB-W	17.5	V
VD	控制电源电压 Control supply voltage	应用于 VP1 – VNC 之间 Applied between VP1-VNC	17.5	V
VIN	输入信号电压 Input signal voltage	应用于 UP, VP, WP, UN ,VN ,WN-VNC之间 Applied between UP, VP, WP, UN, VN, WN-VNC	-1~10	V
VFO	故障输出电压 Fault output voltage	应用于 FO – V <sub>NC</sub> 之间 Applied between FO-VNC	-0.5~VD+0.5	V
IFO	故障输出电流 Fault output current	FO 端子吸入电流值 FO terminal sink current value	1.5	mA
VSC	电流检测端输入电压 Input voltage of current detection terminal	应用于 CIN – V <sub>NC</sub> 之间 Applied between CIN-VNC	-0.5~VD+0.5	V

### 整个系统 Total System

记号 Mark	项目 Project	条件 Condition	额定值 Ratings	单位 Units
VCC(ROT)	电源电压自己保护范围(短路) Power supply voltage self-protection range (short circuit)	$V_D = V_{DB} = 13.5 \sim 16.5V$ $T_j = 150^\circ\text{C}$ , 无重复, 时间小于 2us $T_j = 150^\circ\text{C}$ , no repetition, time is less than 2us	400	V
T <sub>c</sub>	模块正常工作壳体温度 Module working temperature	$-20^\circ\text{C} \leq T_j \leq 150^\circ\text{C}$	-20 ~ +100	°C
T <sub>stg</sub>	贮存温度 Storage temperature	/	-40 ~ +125	°C
V <sub>iso</sub>	绝缘耐压 Insulation withstand voltage	正弦波60Hz, AC 1分钟, 在插脚和散热片之间 Sine wave 60Hz, AC for 1 minute, between pin and heat sink	1500	V <sub>rms</sub>

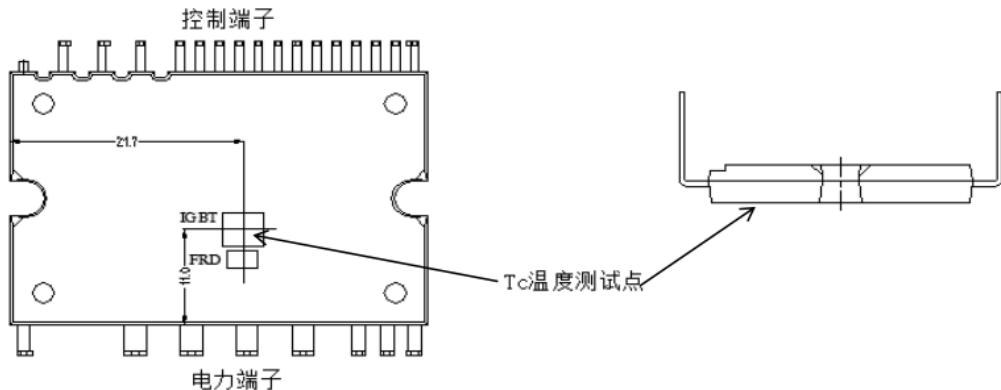
图 4: 壳温  $T_c$  测试点

Fig 4: Case Temperature Measurement

### 热阻 Thermal Resistance

记号 Mark	项目 Project	条件 Condition	最小值 Min.	典型值 Typ.	最大值 Max.	单位 Unit
R <sub>th(j-c)Q</sub>	结点到壳的热阻 Junction to case thermal resistance	单个 IGBT 元件 Single IGBT element	-	-	5.5	°C/W
R <sub>th(j-c)F</sub>		单个 FRD 元件 Single FRD element	-	-	6.2	°C/W

电气特性 ( $T_j=25^\circ\text{C}$ , 除非特殊说明)

**Electrical Characteristics** ( $T_j=25^\circ\text{C}$ , Unless Otherwise Specified)

### 逆变部分 Inverter Part

记号 Mark	项目 Project	条件 Condition	最小值 Min.	典型值 Typ.	最大值 Max.	单位 Unit
$V_{CE(\text{sat})}$	集电极与发射极间饱和电压 Saturation voltage between collector and emitter	$V_D = V_{DB} = 15\text{V}$ $V_{IN} = 5\text{V}, I_C = 6\text{A}, T_j = 25^\circ\text{C}$	-	1.6	2.2	V
		$V_D = V_{DB} = 15\text{V}$ $V_{IN} = 5\text{V}, I_C = 6\text{A}, T_j = 125^\circ\text{C}$	-	1.85	-	V
$V_F$	FWD 正向导通电压 FWD forward voltage	$V_{IN} = 0\text{V}, I_C = -6\text{A}, T_j = 25^\circ\text{C}$		1.4	2.1	V
$t_{ON}$	开关时间 (备注 2) Switching time (Note 2)	$V_{CC} = 300\text{V}, V_D = V_{DB} = 15\text{V}, I_C = 6\text{A}$ $V_{IN} = 0\text{V}-5\text{V}$ , 感性负载	-	620	-	nS
$t_{C(ON)}$			-	95	-	nS
$t_{OFF}$			-	700	-	nS
$t_{C(OFF)}$			-	70	-	nS
$t_{rr}$			-	95	-	nS
$E_{on}$	开通损耗 Turn-on loss	$I_C = 6\text{A}, V_{CC} = 400\text{V}, V_D = V_{DB} = 15\text{V}, L = 1\text{mH}, T_j = 25^\circ\text{C}$	-	42	92	uJ
$E_{off}$	关断损耗 Turn-off loss		-	90	142	





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$I_{CES}$	$V_{CE} = V_{CES}$ $T_j = 25^\circ C$	-	-	75	uA
	$V_{CE} = V_{CES}$ $T_j = 125^\circ C$	-	-	1	mA

备注 2:  $t_{ON}$  和  $t_{OFF}$  包括驱动 IC 内部传输延迟时间。 $t_{C(ON)}$  和  $t_{C(OFF)}$  是 IGBT 自身被内部给定门极驱动条件下的开关时间。详见图 5。

Remark 2:  $t_{ON}$  and  $t_{OFF}$  include the internal transmission delay time of the driver IC.  $t_{C(ON)}$  and  $t_{C(OFF)}$  are the switching times of the IGBT itself driven by the internally given gate. See Figure 5 for details.

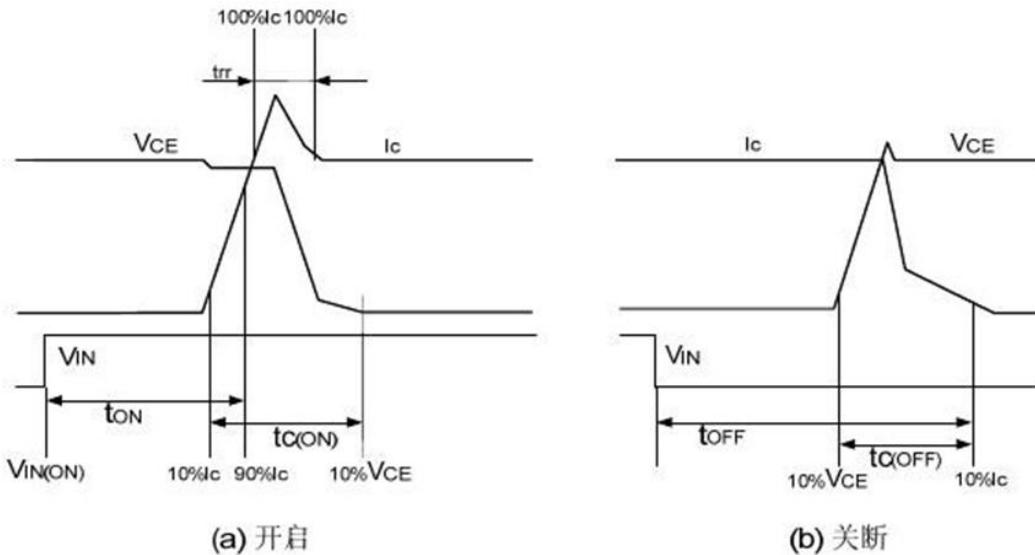


图 5: 开关时间定义

Fig 5: Switching Time Definition

### 控制部分 Control Part

记号 Mark	项目 Project	条件 Condition		最小值 Min.	典型值 Typ.	最大值 Max.	单位 Unit
ID	VD 静态电流 VD Quiescent Current	$VD = 15V$ $VIN = 5V$	VP1-VNC	-	0.35	1	mA
IDB	VDB 静态电流 VDB Quiescent Current	$VDB = 15V$ $VIN = 5V$	UFB - U, VFB - V, WFB - W	-	300	550	uA
$V_{FOH}$	故障输出电压 Fault output voltage	$V_{sc} = 0V$ , $FO$ 脚通过 10K 电阻上拉至 5V $V_{sc} = 0V$ , $FO$ pin is pulled up to 5V through 10K resistor	4.6	-	-	-	V
$V_{FOL}$		$V_{sc} = 1V$ , $I_{FO} = 1.5mA$	-	-	0.3	-	V
$V_{sc,TH}^+$	短路正向触发阈值 Short circuit forward trigger threshold	$V_D = 15V$	0.37	0.47	0.65	-	V
$V_{sc,TH}^-$	短路负向触发阈值 Short circuit negative trigger threshold	$V_D = 15V$	0.2	0.4	-	-	V
$UV_{DD}$	电源欠压保护控制	触发电平 Trigger level	9.5	10.4	11.0	-	V



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UV <sub>DR</sub>	Power undervoltage protection control	复位电平 Reset level	11.0	12.1	12.8	
UVDBD		触发电平 Trigger level	9.5	10.4	11.0	
UVDBR		复位电平 Reset level	11.0	12.1	12.8	
Ron,FLT	故障低有效阻抗 Fault low effective impedance	I=1.5mA		50	90	ohm
T <sub>FO</sub>	故障输出脉冲宽度 Fault output pulse width	-	40	76	120	uS
t <sub>FIL,IN</sub>	输入信号滤过时间 (UP/VP/WP, UN/VN/WN) Input signal filtering time	V <sub>IN</sub> = 0 V & 5 V	140	290		nS
t <sub>CINMIN</sub>	CIN 输入信号滤过时间 CIN input signal filtering time	V <sub>IN</sub> = 0 V or 5 V, V <sub>CIN</sub> = 5 V	270	530	780	nS
V <sub>IN(ON)</sub>	开启阙值电压 Turn on threshold voltage	应用于 UP, VP, WP, UN, VN, WN 和 VNC 之间 Applied between UP, VP, WP, UN, VN, WN and VNC	1.6	2.1	2.5	V
V <sub>IN(OFF)</sub>	关断阙值电压 Turn-off threshold voltage		0.8	0.94	1.2	
VOT	温度输出, 备注 3 Temperature output, remark 3	T <sub>c</sub> =90°C	1.53	1.59	1.65	V
		T <sub>c</sub> =25°C	4.15	4.17	4.19	
V <sub>F</sub>	BSD 正向电压 BSD forward voltage	I <sub>F</sub> =10mA 包含电压 I <sub>F</sub> =10mA Including voltage	-	3.25	5	V
R <sub>BSD</sub>	BSD限流电阻 BSD current limiting resistor	VF1=4V, VF2=5V	18	30	42	ohm

备注 3: 当温度达到极限时, IPM 不能自动够关闭 IGBT 和输出故障信号。当温度超出使用者定义的限定值时, 应使用控制器(单片机)关闭 IPM。IPM 的 VOT 输出特性曲线请参考图 6, 图 6 曲线是以 20K 上拉电阻测试结果。

Remark 3: When the temperature reaches the limit, the IPM cannot automatically turn off the IGBT and output a fault signal. When the temperature exceeds the user-defined limit value, the controller (microcontroller) should be used to turn off the IPM. Please refer to Figure 6 for the VOT output characteristic curve of IPM. The curve in Figure 6 is the test result of 20K pull-up resistor.

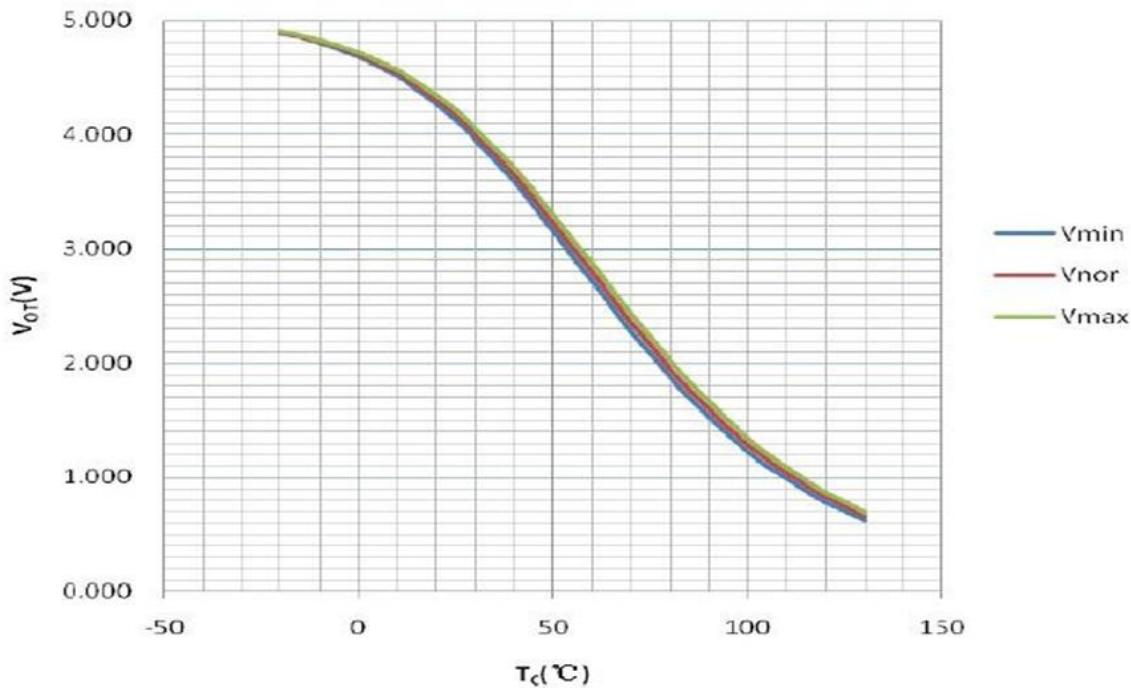


图 6: VOT 输出电压曲线

Figure 6: VOT output voltage curve

### 推荐工作条件 Recommended Operating Conditions

记号 Mark	项目 Project	条件 Condition	最小值 Min.	典型值 Typ.	最大值 Max.	单位 Unit
Vcc	电源电压 voltage	应用于 P – NU, NV, NW 之间 Applied between P – NU, NV, NW	0	300	400	V
VD	控制电源电压 Control supply voltage	应用于 VP1 – VNC 之间 Applied between VP1-VNC	-	15	-	V
VDS	上臂控制电源电压 Upper arm control power supply voltage	应用于 VUFB – U, VVFB – V, VWFB-W 之间 Applied between VUFB-U, VVFB-V, VWFB-W	-	15	-	V
tdead	死区时间 Dead time	各桥臂输入对应, T <sub>c</sub> <=100°C Corresponding to the input of each bridge arm, T <sub>c</sub> <=100°C	1	-	-	us
fPWM	PWM 频率 PWM frequency	-20°C≤T <sub>c</sub> ≤+100°C -20°C≤T <sub>j</sub> ≤+150°C	-	-	20	kHz
PWM	最小输入信号脉冲宽度 Minimum input signal pulse width	ON	0.7	-	-	us
		OFF	0.7	-	-	us
T <sub>j</sub>	结温 Junction temperature	-	-20	-	125	°C

**内部 NTC –热阻特性 Internal NTC-thermal resistance characteristics**

记号 Mark	项目 Project	条件 Condition	最小值 Min.	典型值 Typ.	最大值 Max.	单位 Unit
RNTC	热敏电阻 Thermistor	TNTC = 25°C	97	100	103	Kohm
		TNTC = 125°C	3.25	3.46	3.69	Kohm
温度范围 temperature range			-40	-	+125	°C

**机械特性 Mechanical properties**

项目 Project	条件 Condition	最小值 Min.	典型值 Typ.	最大值 Max.	单位 Unit
安装扭矩 Installation torque	螺丝钉尺寸: M3 Screw size: M3	-	0.69	-	N·m
设计平面度 Design flatness	见图 7 Figure 7	-50	-	+120	um
重量 weight	-	-	7	-	g

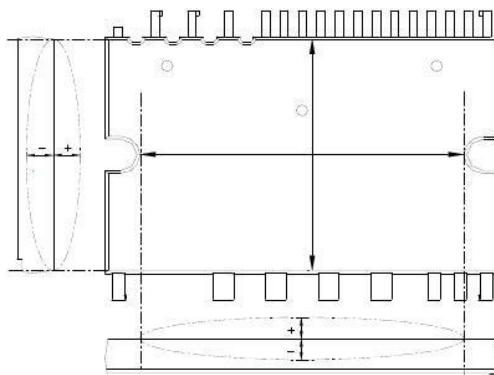


图 7: 平面度检测位置

Figure 7: Flatness detection position

**应用指南 Application Guide****增强型输入滤波 Enhanced input filtering**

增强型滤波器能够改善 HVIC 内部模块的输入/输出脉冲的一致性及有助于滤除尖峰干扰信号和窄脉冲，如下图 8 和图 9 是经典型输入滤波器和增强型输入滤波器演示图。

The enhanced filter can improve the consistency of the input/output pulses of the internal modules of the HVIC and help to filter out the peak interference signals and narrow pulses, as shown in Figures 8 and 9 below are the classic input filter and enhanced input filter demonstration diagram.

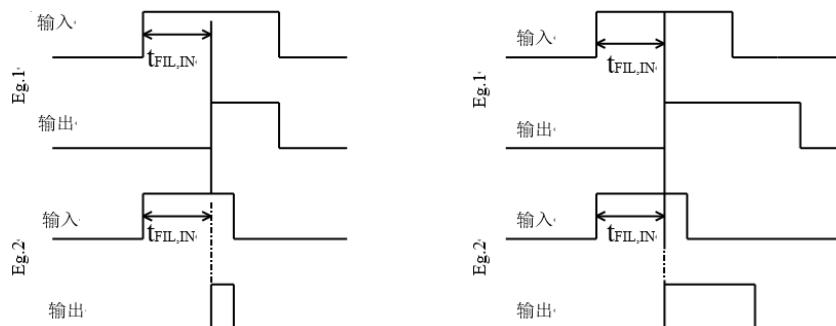


图 8: 典型输入滤波

Figure 8: Typical input filtering Figure

图 9: 增强型输入滤波

9: Enhanced input filtering

## 保护功能时序图 Timing diagram of protection function

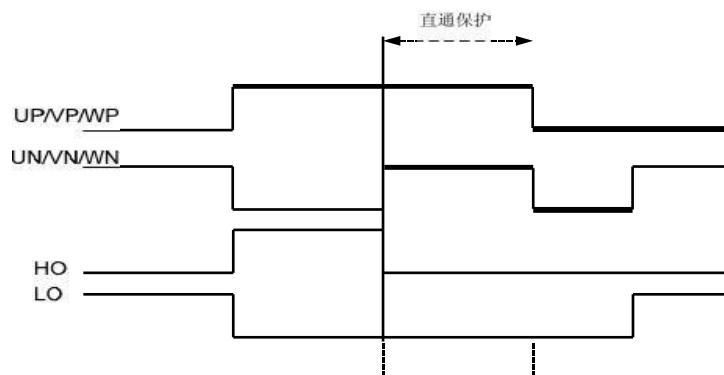


图 10: 直通保护

Figure 10: Pass-through protection

备注 4: HO 和 LO 为内部 HVIC 门极输出信号。

Note 4: HO and LO are internal HVIC gate output signals.

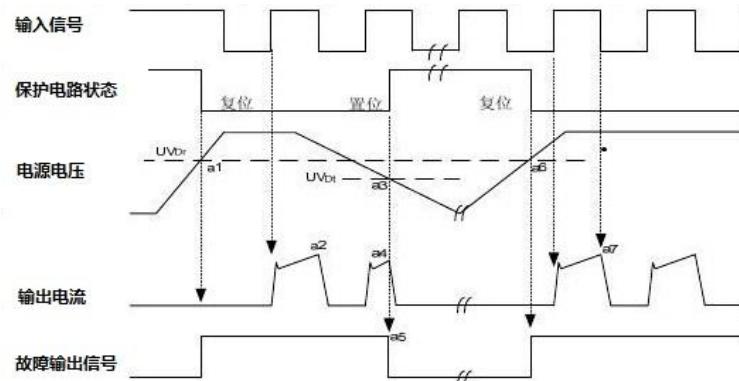


图 11: 欠压保护时序图(低侧)

Figure 11: Timing diagram of undervoltage protection (low side)

a1: 电源电压上升: 当该电压上升到欠压恢复点, 在下一个欠压信号被执行前该线路将启动运行。

a1: Power supply voltage rise: When the voltage rises to the undervoltage recovery point, the line will start running before the next undervoltage signal is executed.

a2: 正常运行: IGBT 开启并加载电流。

a2: Normal operation: IGBT turns on and loads current.

a3: 欠压检测点(UVDt)。

a3: Undervoltage detection point (UVDt).

a4: 不管输入是什么信号, IGBT 都是关闭状态。  
a4: No matter what signal is input, the IGBT is off.

a5: 故障输出开启。

a5: Fault output is on.

a6: 欠压恢复(UVDr)。

a6: Undervoltage recovery (UVDr).

a7: 正常运行: IGBT 导通并加载负载电流。

a7: Normal operation: IGBT is turned on and load current is loaded.

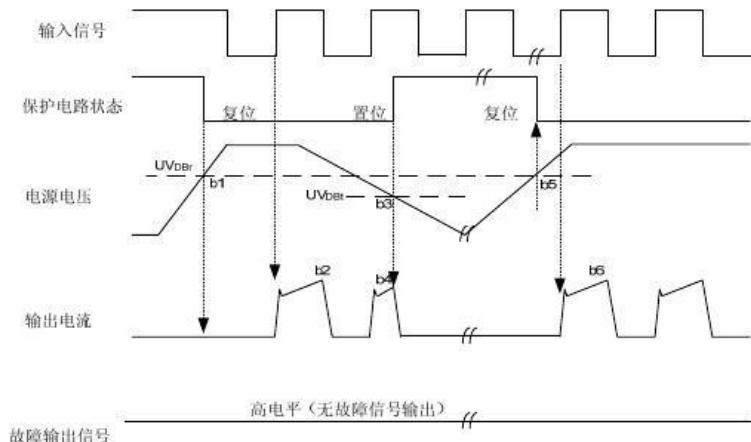


图 12: 欠压保护时序图 (高侧)  
Figure 12: Timing diagram of undervoltage protection (high side)

b1 : 电源电压上升: 当该电压上升到欠压恢复点, 在下一个欠压信号被执行前该线路将启动运行。

b1: Power supply voltage rise: When the voltage rises to the undervoltage recovery point, the line will start running before the next undervoltage signal is executed.

b2 : 正常运行: IGBT 导通并加载负载电流。

b2: Normal operation: IGBT is turned on and load current is applied.

b3 : 欠压检测 (UVDBt)。

b3: Undervoltage detection (UVDBt).

b4 : 不管输入是什么信号, IGBT 都是关闭状态。

b4: No matter what signal is input, IGBT is off.

b5 : 欠压恢复(UVDBr)。

b5: Undervoltage recovery (UVDBr).

b6 : 正常运行: IGBT 导通并加载负载电流。

b6: Normal operation: IGBT is turned on and load current is applied.

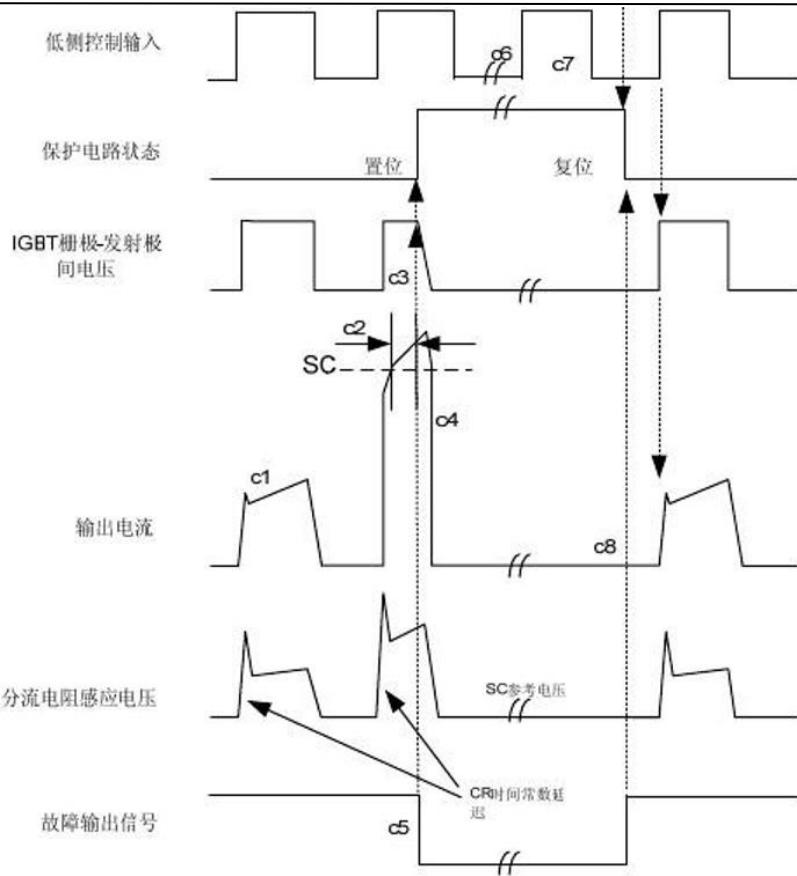


图 13: 短路电流保护时序图 (只适合于低侧)

(通过外部分流电阻连接)

Figure 13: Timing diagram of short-circuit current protection (only suitable for low side)

(Connected via external shunt resistor)

c1: 正常运行: IGBT 导通载流。

c1: Normal operation: IGBT conducting current.

c2: 短路电流检测 (CIN 触发器)。

c2: Short-circuit current detection (CIN trigger).

c3: IGBT 门极被强制关断。

c3: IGBT gate is forcibly turned off. c4: IGBT is turned off.

c4 : IGBT 关断。

c4: IGBT is turned off.

c5: 故障输出定时器开始运行: 故障输出信号的脉冲宽度是由外部电容 CFO 设定。

c5: The fault output timer starts to run: The pulse width of the fault output signal is set by the external capacitor CFO.

c6: 输入“L”: IGBT 关闭。

c6: Input "L": IGBT is off.

c7: 输入 “H”: IGBT 开通, 但是故障信号作用期间, IGBT 不导通。

c7: Input "H": The IGBT is turned on, but the IGBT is not turned on during the fault signal.

c8: IGBT 关断。

c8: IGBT is turned off.

## 输入/输出接口电路 Input/output interface circuit

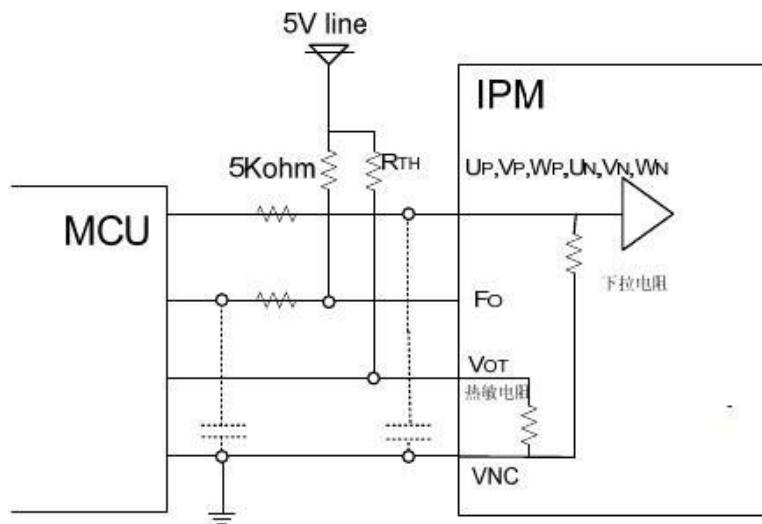


图 14: 推荐的 MCU 输入输出接口电路

Figure 14: Recommended MCU input and output interface circuit

备注 5: 由于 PWM 的控制方式和实际应用电路的阻抗及线路板的阻抗, RC 去耦可能会有变化。

Note 5: Due to the PWM control method and the impedance of the actual application circuit and the impedance of the circuit board, RC decoupling may change.

备注 6: 逻辑输入要和标准的 CMOS 或 LSTTL 输出相匹配。

Note 6: The logic input must match the standard CMOS or LSTTL output.

## 分流电阻接线 Shunt resistor wiring

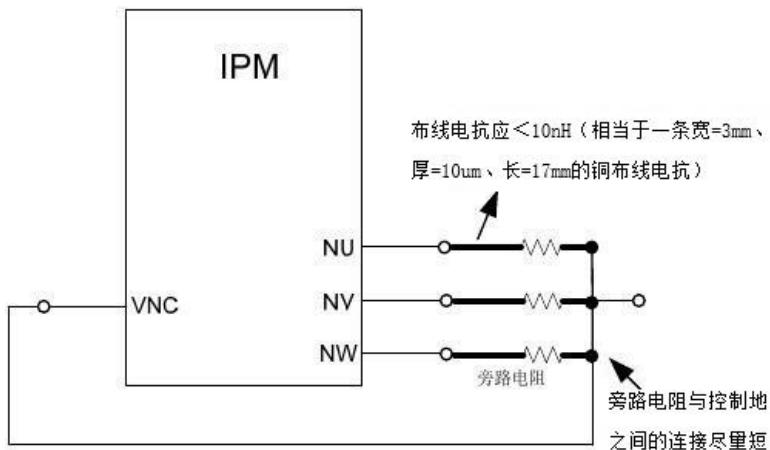


图 15: 旁路电阻接线注意事项

Figure 15: Precautions for bypass resistance wiring

## 典型应用电路图 Typical application circuit diagram

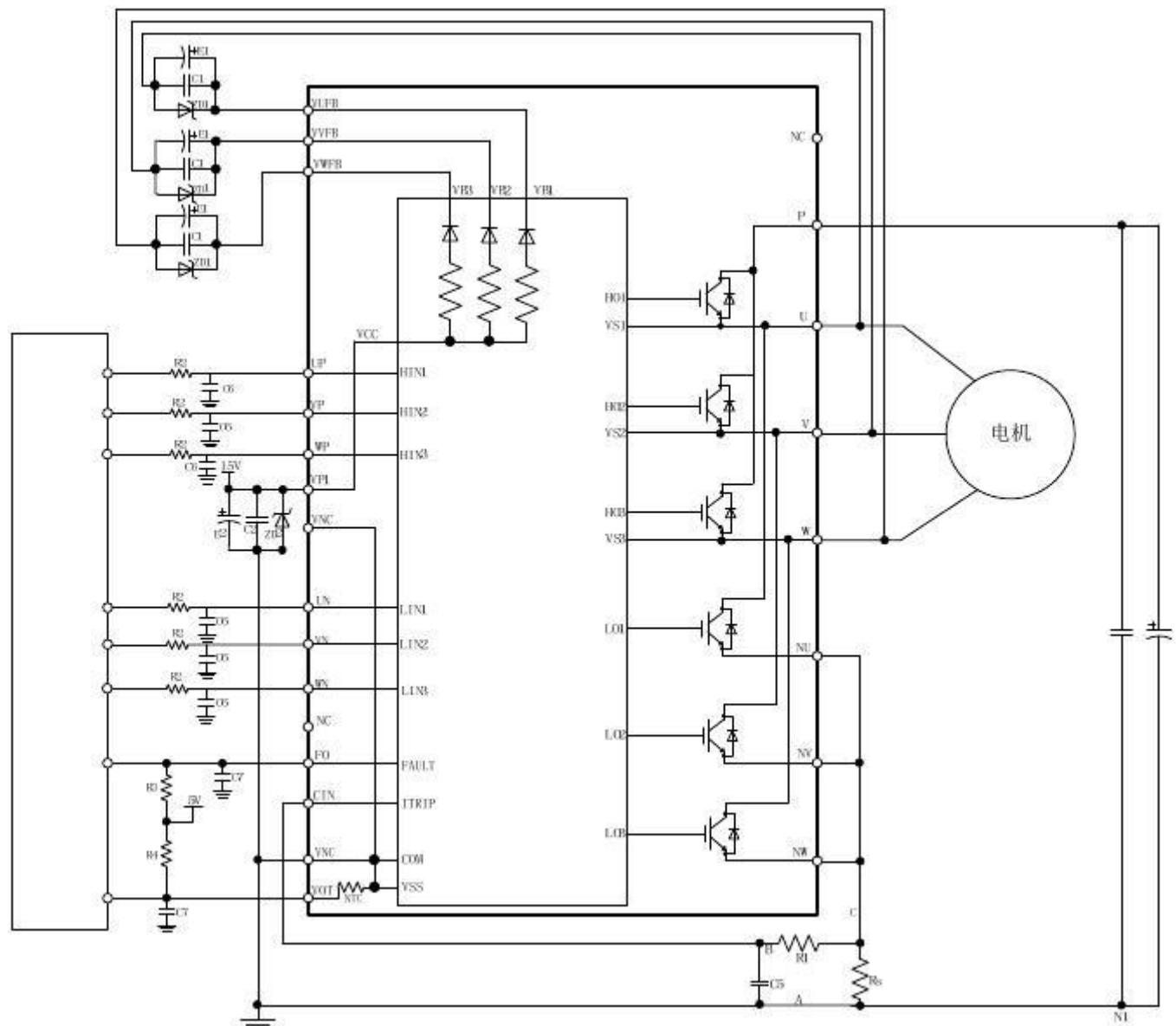


图 16: 典型应用电路

Figure 16: Typical application circuit

备注 7: 输入驱动高有效; IC 内部集成有一个  $5k\Omega$ (典型值)下拉电阻; 为防止发生误动作, 输入布线应尽可能短; 当用 RC 去耦线路时, 须确保输入信号达到开启和关断阈值电压范围。

Remark 7: The input drive is highly effective; a  $5k\Omega$  (typical value) pull-down resistor is integrated inside the IC; in order to prevent malfunctions, the input wiring should be as short as possible; when using RC decoupling lines, make sure that the input signal reaches on and off Threshold voltage range.

备注 8: 由于模块内置了专用 HVIC, 其控制端子可与 CPU 端子直接相连, 而不需要任何光耦或变压器等隔离电路。



Remark 8: Since the module has a built-in dedicated HVIC, its control terminal can be directly connected to the CPU terminal without any isolation circuit such as optocoupler or transformer.

备注 9: 自举电路负极应直接连接到 U、V、W 的端。

Remark 9: The negative pole of the bootstrap circuit should be directly connected to the U, V, W terminals.

备注 10: FO 是漏极开路型，其信号线应通过一个约  $10k\Omega$  的上拉电阻上拉到+5V/3.3V 电源。

Remark 10: FO is an open-drain type, and its signal line should be pulled up to a +5V/3.3V power supply through a pull-up resistor of about  $10k\Omega$ .

备注 11: 为防止误保护，A、B、C 连线应尽可能短。

Note 11: To prevent false protection, the A, B, and C wiring should be as short as possible.

备注 12: 保护线路 R1、C5 的时间常数建议选取在 1~2uS。关断时间可能随着布线的不同而多少有些变化。建议 R1、C5 选择小容差，温度补偿类型。

Remark 12: The time constant of protection lines R1 and C5 is recommended to be 1~2uS. The turn-off time may vary somewhat depending on the wiring. It is recommended that R1 and C5 choose small tolerance and temperature compensation type.

备注 13: 所有电容的位置尽可能的靠近 IPM。

Remark 13: Position all capacitors as close to IPM as possible.

备注 14: 为了防止噪声干扰，储能电容与 P&N1 之间的引线应尽可能的短，推荐在 P&N1 端子之间加约 0.1~0.22uF 的 MLCC 低频滤波电容。

Note 14: In order to prevent noise interference, the lead between the storage capacitor and P&N1 should be as short as possible. It is recommended to add about 0.1~0.22uF of MLCC low-frequency filter capacitor between P&N1 terminals.

备注 15: 两个 VNC 端（9&16 脚）在 IPM 内部已连接在一起，外部任一 VNC 端子连接到 GND，另一端子允许开路。

Remark 15: The two VNC terminals (pins 9 & 16) are connected together inside the IPM, any external VNC terminal is connected to GND, and the other terminal is allowed to open.

备注 16: 如果控制地通过 PCB 走线被连接到功率地，控制信号可能会受到功率地的起伏的影响，推荐使用单点连接。

Remark 16: If the control ground is connected to the power ground through PCB traces, the control signal may be affected by the fluctuation of the power ground. It is recommended to use a single point connection.

## 外形封装图 Outline package drawing

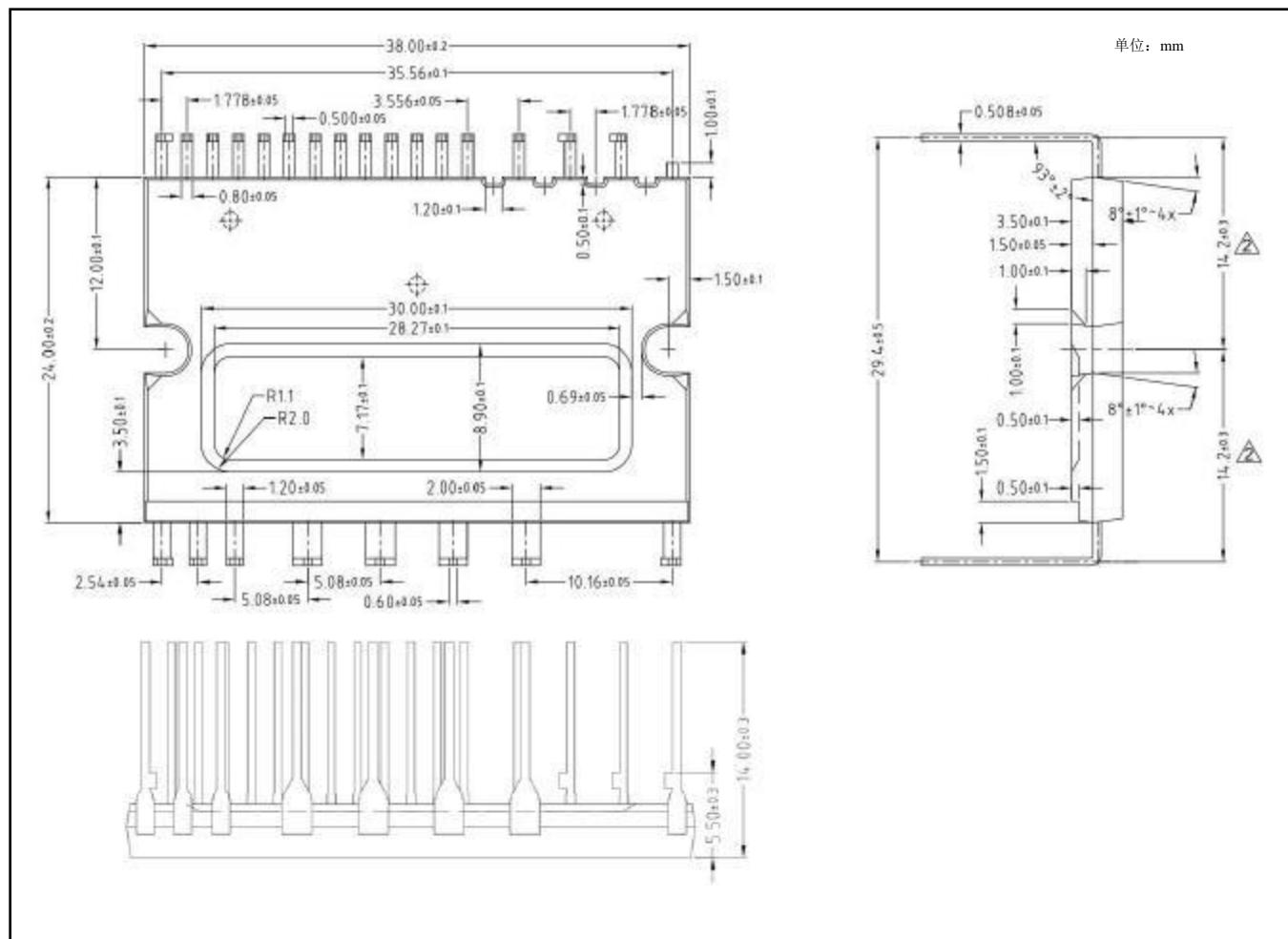


图 17：封装尺寸示意图

Figure 17: Schematic diagram of package size



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